## Accurate Evaluation of Characteristic Modes

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## Outline

(1) Characteristic Modes
(2) Definition of Matrix $\mathbf{S}$
(3) Modification of Generalized Eigenvalue Problem
(4) Decomposition With Matrix $\mathbf{S}$
(5) Other Applications
(6) Concluding Remarks

$\mathcal{J}_{2}(\boldsymbol{r}, t)$

This talk concerns:

- electric currents in vacuum (generalization is, however, straightforward),
- time-harmonic quantities, i.e., $\boldsymbol{\mathcal { A }}(\boldsymbol{r}, t)=\operatorname{Re}\{\boldsymbol{A}(\boldsymbol{r}) \exp (\mathrm{j} \omega t)\}$.


## Characteristic Mode Decomposition

Generalized eigenvalue problem ${ }^{1}$

$$
\mathbf{X} \mathbf{I}_{n}=\lambda_{n} \mathbf{R} \mathbf{I}_{n}
$$

$\mathbf{Z}=\mathbf{R}+\mathrm{j} \mathbf{X} \in \mathbb{C}^{N \times N}$ is impedance matrix, $\mathbf{I}_{n} \in \mathbb{R}^{N \times 1}$ are expansion coefficients.

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- provide physical insight
- formalization of what antenna designers know and understand
- excellent entire-domain basis

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- hyped and sometimes misused (since used for everything)
- suffers from numerical problems
- incompatible with realistic feeding

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[^3]
## Benchmark of CM Solvers: Spherical Shell ${ }^{2}$, $k a=1 / 2$



[^4]
## Cause of Limited Number of Modes

Previous benchmark generated some important questions:

- How many modes can, in principle, be found?
- Is there a way how to increase their number?
- Is there a way how to accelerate solution if only few modes are needed?



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- Is there a way how to accelerate solution if only few modes are needed?

Problem is predominantly caused by numerical dynamics of the $\mathbf{R}$ matrix (naive interpretation: only a few modes radiate well. You will see later...).


## Electric Field Integral Equation (EFIE)

EFIE for PEC bodies as the core of underlying MoM formulation:

$$
\begin{equation*}
\hat{\boldsymbol{n}} \times \boldsymbol{E}\left(\boldsymbol{r}_{2}\right)=\mathrm{j} k Z_{0} \hat{\boldsymbol{n}} \times \int_{\Omega} \mathbf{G}\left(\boldsymbol{r}_{1}, \boldsymbol{r}_{2}\right) \cdot \boldsymbol{J}\left(\boldsymbol{r}_{1}\right) \mathrm{d} S_{1}, \tag{1}
\end{equation*}
$$

with dyadic Green function defined as

$$
\begin{equation*}
\mathbf{G}\left(\boldsymbol{r}_{1}, \boldsymbol{r}_{2}\right)=\left(\mathbf{1}+\frac{1}{k^{2}} \nabla \nabla\right) \frac{\mathrm{e}^{-\mathrm{j} k\left|\boldsymbol{r}_{1}-\boldsymbol{r}_{2}\right|}}{4 \pi\left|\boldsymbol{r}_{1}-\boldsymbol{r}_{2}\right|} \tag{2}
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\end{equation*}
$$

The impedance matrix $\mathbf{Z}$ reads

$$
\begin{equation*}
Z_{p q}=\mathrm{j} k Z_{0} \int_{\Omega} \int_{\Omega} \boldsymbol{\psi}_{p}\left(\boldsymbol{r}_{1}\right) \cdot \mathbf{G}\left(\boldsymbol{r}_{1}, \boldsymbol{r}_{2}\right) \cdot \boldsymbol{\psi}_{q}\left(\boldsymbol{r}_{2}\right) \mathrm{d} S_{1} \mathrm{~d} S_{2} \tag{3}
\end{equation*}
$$

## Spherical Wave Expansion of Dyadic Green Function

Spherical wave expansion of dyadic Green function reads ${ }^{3}$

$$
\begin{equation*}
\mathbf{G}\left(\boldsymbol{r}_{1}, \boldsymbol{r}_{2}\right)=-\mathrm{j} k \sum_{\alpha} \mathbf{u}_{\alpha}^{(1)}\left(k \boldsymbol{r}_{<}\right) \mathbf{u}_{\alpha}^{(4)}\left(k \boldsymbol{r}_{>}\right) . \tag{4}
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Impedance matrix $\mathbf{Z}$ with spherical wave expansion substituted

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\begin{equation*}
Z_{p q}=k^{2} Z_{0} \sum_{\alpha} \int_{\Omega} \int_{\Omega} \boldsymbol{\psi}_{p}\left(\boldsymbol{r}_{1}\right) \cdot \mathbf{u}_{\alpha}^{(1)}\left(k \boldsymbol{r}_{<}\right) \mathbf{u}_{\alpha}^{(4)}\left(k \boldsymbol{r}_{>}\right) \cdot \boldsymbol{\psi}_{q}\left(\boldsymbol{r}_{2}\right) \mathrm{d} S_{1} \mathrm{~d} S_{2} \tag{5}
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\end{equation*}
$$

can be used for reformulation of matrix $\mathbf{R}$ since $\mathbf{u}_{\alpha}^{(1)}(k \boldsymbol{r})=\operatorname{Re}\left\{\mathbf{u}_{\alpha}^{(4)}(k \boldsymbol{r})\right\}$ as

$$
\begin{equation*}
R_{p q}=k^{2} Z_{0} \sum_{\alpha} \int_{\Omega} \boldsymbol{\psi}_{p}\left(\boldsymbol{r}_{1}\right) \cdot \mathbf{u}_{\alpha}^{(1)}\left(k \boldsymbol{r}_{1}\right) \mathrm{d} S_{1} \int_{\Omega} \mathbf{u}_{\alpha}^{(1)}\left(k \boldsymbol{r}_{2}\right) \cdot \boldsymbol{\psi}_{q}\left(\boldsymbol{r}_{2}\right) \mathrm{d} S_{2} \tag{6}
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$$

[^8]
## Definition of Projection Matrix S

Resistance matrix $\mathbf{R}$ is expressed as a product of two identical rectangular matrices:

$$
R_{p q}=\sum_{\alpha}\left(k \sqrt{Z_{0}} \int_{\Omega} \boldsymbol{\psi}_{p}\left(\boldsymbol{r}_{1}\right) \cdot \mathbf{u}_{\alpha}^{(1)}\left(k \boldsymbol{r}_{1}\right) \mathrm{d} S_{1}\right)\left(k \sqrt{Z_{0}} \int_{\Omega} \mathbf{u}_{\alpha}^{(1)}\left(k \boldsymbol{r}_{2}\right) \cdot \boldsymbol{\psi}_{q}\left(\boldsymbol{r}_{2}\right) \mathrm{d} S_{2}\right)
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$$

Definition ${ }^{4}$ of the matrix $\mathrm{S} \in \mathbb{R}^{N_{\alpha} \times N}$

$$
\begin{equation*}
S_{\alpha p}=k \sqrt{Z_{0}} \int_{\Omega} \boldsymbol{\psi}_{p}(\boldsymbol{r}) \cdot \mathbf{u}_{\alpha}^{(1)}(k \boldsymbol{r}) \mathrm{d} S, \tag{6}
\end{equation*}
$$

and its relation to the resistance matrix

$$
\begin{equation*}
\mathbf{R}=\mathbf{S}^{\mathrm{T}} \mathbf{S} \tag{7}
\end{equation*}
$$

|  | S |
| :---: | :---: |
| $\mathbf{S}^{\text {T }}$ | $\mathbf{R}=\mathbf{S}^{\mathrm{T}} \mathbf{S}$ |

[^9]Properties of Matrix S, Part \#1


## Properties of Matrix S, Part \#1

- Matrix $\mathbf{S}$ is real-valued, rectangular, low-rank

$$
\begin{align*}
N_{\alpha} & =2 L(L+2),  \tag{8}\\
L & =\lceil k a+7 \sqrt[3]{k a}+2\rceil . \tag{9}
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- Matrix $\mathbf{S}^{\mathrm{T}} \mathbf{S}$ does not contain any negative eigenvalue higher than numerical noise.
- Matrix $\mathbf{S}$ represents projection between RWGs and spherical waves, i.e.,

$$
\begin{align*}
\mathbf{R} & =\mathbf{S}^{\mathrm{T}} \mathbf{S}  \tag{10}\\
\mathbf{R}^{\mathrm{sph}} & =\mathbf{S S}^{\mathrm{T}} \tag{11}
\end{align*}
$$



## Properties of Matrix S, Part \#2

Radiated power can be calculated as

$$
\begin{equation*}
P_{\mathrm{rad}}=\frac{1}{2 Z_{0}} \int_{S^{2}}|\boldsymbol{F}(\hat{\boldsymbol{r}})|^{2} \mathrm{~d} S \approx \frac{1}{2} \mathbf{I}^{\mathrm{H}} \mathbf{R I}=\frac{1}{2}|\mathbf{S I}|^{2}=\frac{1}{2} \sum_{\alpha}\left|f_{\alpha}\right|^{2} \tag{12}
\end{equation*}
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with

$$
\begin{equation*}
\boldsymbol{F}(\hat{\boldsymbol{r}})=\frac{1}{k} \sum_{\alpha} \mathrm{j}^{l-\tau+2} f_{\alpha} \mathbf{Y}_{\alpha}(\hat{\boldsymbol{r}}), \tag{13}
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where $\mathbf{Y}_{\alpha}(\hat{\boldsymbol{r}})$ are the real-valued spherical vector harmonics.

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| Example |  |  |  | Comp. times in IDA (s) |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $N_{\alpha}$ | $N$ | $\mathbf{R}$ | $\mathbf{S}$ | $\mathbf{R}=\mathbf{S}^{\mathrm{T}} \mathbf{S}$ |  |
| spherical shell | 880 | 750 | 0.09 | 0.009 | 0.011 |  |
| spherical shell | 880 | 3330 | 1.78 | 0.039 | 0.083 |  |
| helicopter | 880 | 18898 | 54.50 | 0.236 | 1.660 |  |

## CMs Using SVD of matrix $\mathbf{S}$ and GEP Partitioning

Singular value decomposition (SVD) of matrix $\mathbf{S}$

$$
\begin{equation*}
\mathbf{S}=\mathbf{U} \boldsymbol{\Lambda} \mathbf{V}^{\mathrm{H}} \tag{14}
\end{equation*}
$$

substituted into CM definition gives

$$
\begin{equation*}
\left(\mathbf{V}^{\mathrm{H}} \mathbf{X V}\right)\left(\mathbf{V}^{\mathrm{H}} \mathbf{I}_{n}\right)=\lambda_{n}\left(\boldsymbol{\Lambda}^{\mathrm{H}} \boldsymbol{\Lambda}\right)\left(\mathbf{V}^{\mathrm{H}} \mathbf{I}_{n}\right) \quad \longrightarrow \quad \widetilde{\mathbf{X}} \widetilde{\mathbf{I}}_{n}=\lambda_{n} \widetilde{\mathbf{R}} \widetilde{\mathbf{I}}_{n}, \tag{15}
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$$

Partitioning

$$
\widetilde{\mathbf{X}} \widetilde{\mathbf{I}}=\left(\begin{array}{ll}
\widetilde{\mathbf{X}}_{11} & \widetilde{\mathbf{X}}_{12}  \tag{16}\\
\widetilde{\mathbf{X}}_{21} & \widetilde{\mathbf{X}}_{22}
\end{array}\right)\binom{\widetilde{\mathbf{I}}_{1 n}}{\widetilde{\mathbf{I}}_{2 n}}=\binom{\lambda_{1 n} \widetilde{\mathbf{R}}_{11} \widetilde{\mathbf{I}}_{1 n}}{\mathbf{0}}
$$

and reducing to Schur complement yields the final GEP formulation

$$
\begin{equation*}
\left(\widetilde{\mathbf{X}}_{11}-\widetilde{\mathbf{X}}_{12} \widetilde{\mathbf{X}}_{22}^{-1} \widetilde{\mathbf{X}}_{21}\right) \widetilde{\mathbf{I}}_{1 n}=\lambda_{1 n} \widetilde{\mathbf{R}}_{11} \widetilde{\mathbf{I}}_{1 n} . \tag{17}
\end{equation*}
$$

Properties of the Decomposition:

Characteristic modes are constructed as

$$
\begin{equation*}
\widetilde{\mathbf{I}}_{n}=\binom{\widetilde{\mathbf{I}}_{1 n}}{-\widetilde{\mathbf{X}}_{22}^{-1} \widetilde{\mathbf{X}}_{21} \widetilde{\mathbf{I}}_{1 n}}, \tag{18}
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$$

radiated power is implicitly normalized by $\boldsymbol{\Lambda}^{\mathrm{H}} \boldsymbol{\Lambda}$ matrix in (15)

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\begin{equation*}
\widetilde{\mathbf{I}}_{n}^{\mathrm{H}} \widetilde{\mathbf{R}} \widetilde{\mathbf{I}}_{m}=\delta_{n m} \tag{19}
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Properties ${ }^{4}$ :

- numerical dynamics doubled thanks to the SVD and partitioning,
- number of used spherical modes controls the number of CMs,
- for $N_{\alpha} \ll N$ (always fulfilled in ESA regime) remarkable speed-up.

[^11]
## Spherical Shell



## Spherical Shell



## Spherical Shell



## Rectangular Plate

100 modes were calculated (eigs)

- (X, R) 0.7s (29)
- (X, R) + Advanpix: 1324 s
- ( $\widetilde{\mathbf{X}}, \widetilde{\mathbf{R}}) 0.5 \mathrm{~s}(37)$
(If matrix $\mathbf{S}$ is reduced, calculation further accelerated.)

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## Rectangular Plate - Higher-order Modes

Two high-order modes of rectangular plate:

- left: inductive, $n=17$,
$\lambda_{17}=2.461 \cdot 10^{17}$,
- right: capacitive, $n=77$, $\lambda_{77}=-1.947 \cdot 10^{24}$.



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- left: inductive, $n=17$,
$\lambda_{17}=2.461 \cdot 10^{17}$,
- right: capacitive, $n=77$, $\lambda_{77}=-1.947 \cdot 10^{24}$.

Such high-order modes are not needed in practice (except tracking).

- However, accuracy can be interchanged for comp. speed.



## Acceleration of the CMs Decomposition

If double precision is enough, however, computational speed is required:

$$
\begin{equation*}
\mathbf{X} \mathbf{I}_{n}=\lambda_{n} \mathbf{S}^{\mathrm{T}} \mathbf{S} \mathbf{I}_{n} \tag{20}
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$$
\begin{equation*}
\mathbf{S I}_{n}=\lambda_{n} \mathbf{S X}^{-1} \mathbf{S}^{\mathrm{T}} \mathbf{S} \mathbf{I}_{n} \quad \longrightarrow \widehat{\mathbf{X}} \widehat{\mathbf{I}}_{n}=\xi_{n} \widehat{\mathbf{I}}_{n} \tag{21}
\end{equation*}
$$

with $\widehat{\mathbf{X}}=\mathbf{S X} \mathbf{X}^{-1} \mathbf{S}^{\mathrm{T}}, \widehat{\mathbf{I}}_{n}=\mathbf{S I}$, and $\xi_{n}=1 / \lambda_{n}$.

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Properties:

- solved in basis of spherical waves $\left(\widehat{\mathbf{I}}_{n}=\mathbf{S I}\right)$,
- standard (not generalized) eigenvalue problem,
- solution of typically small $N_{\alpha} \times N_{\alpha}$ eigenvalue problem (extreme speed-up),
- all modes, well-represented in the spherical basis, are found,
- eig shall be used instead of eigs in MATLAB.


## Acceleration of the CMs Decomposition - Comparison

| Example |  |  |  |  | Comp. times (s) |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | CMs | $N_{\alpha}$ | $N$ | $(\mathbf{R}, \mathbf{X})$ | $(\widetilde{\mathbf{R}}, \widetilde{\mathbf{X}})$ | $\left(\mathbf{S X}^{-1} \mathbf{S}^{\mathrm{T}}\right)$ |  |
| rectangular plate | 100 | 510 | 655 | 0.7 | 0.8 | 0.5 | $(510$ modes $)$ |
| spherical shell | 300 | 880 | 3330 | $\mathbf{2 9}$ | 6.7 | $\mathbf{2 . 6}$ | (880 modes) |
| helicopter | 25 | 880 | 18898 | $\mathbf{1 4 9}$ | 170 | $\mathbf{4 7}$ | (880 modes) |
| helicopter | 100 | 880 | 18898 | $\mathbf{4 7 3}$ | 173 | $\mathbf{4 7}$ | $(880$ modes $)$ |

Windows Server 2012, $2 \times$ XEON E5-2665 @ 2.4 GHZ, 72 GB RAM

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- ( $\widetilde{\mathbf{R}}, \widetilde{\mathbf{X}})$ gives significantly more modes accurately and is typically faster.
- $\left(\mathbf{S X}^{-1} \mathbf{S}^{\mathrm{T}}\right)$ gives slightly more modes accurately and is significantly faster.
- $\left(\mathbf{S X}^{-1} \mathbf{S}^{\mathrm{T}}\right)$ finds all modes available from a given set of spherical harmonics.
- ( $\left.\mathbf{S X}^{-1} \mathbf{S}^{\mathrm{T}}\right)$ decomposition is excellent for high $k a$ with large DOFs $N$.


## Restriction to TM/TE Modes

Matrix $\mathbf{S}^{\mathrm{TM} / \mathrm{TE}}=\mathbf{S}(i,:)$ contains TE and TM modes in separate rows.

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## Conclusions

- New matrix operator based on MoM formalism,
- matrix $\mathbf{S}$ has controllable and predictable behavior and numerically neat properties,
- matrix $\mathbf{S}$ has many applications (some of them probably yet unknown),
- if $\mathbf{X}$ is not needed, matrix $\mathbf{S}$ should be preferred over $\mathbf{R}$,
- with respect to the (characteristic) modes, the matrix $\mathbf{S}$ is, in certain sense,


Dominant characteristic mode of helicopter model discretized into 18989 RWGs, $k a=1 / 2$. a return to their scattering origin ${ }^{5}$.

[^12]
## Questions?

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[^0]:    ${ }^{1}$ R. F. Harrington and J. R. Mautz, "Computation of characteristic modes for conducting bodies", IEEE Trans. Antennas Propag., vol. 19, no. 5, pp. 629-639, 1971. DoI: 10.1109/TAP.1971.1139990

[^1]:    ${ }^{1}$ R. F. Harrington and J. R. Mautz, "Computation of characteristic modes for conducting bodies", IEEE Trans. Antennas Propag., vol. 19, no. 5, pp. 629-639, 1971. DOI: 10.1109/TAP.1971.1139990

[^2]:    ${ }^{1}$ R. F. Harrington and J. R. Mautz, "Computation of characteristic modes for conducting bodies", IEEE Trans. Antennas Propag., vol. 19, no. 5, pp. 629-639, 1971. DOI: 10.1109/TAP.1971.1139990

[^3]:    ${ }^{1}$ R. F. Harrington and J. R. Mautz, "Computation of characteristic modes for conducting bodies", IEEE Trans. Antennas Propag., vol. 19, no. 5, pp. 629-639, 1971. DOI: 10.1109/TAP.1971.1139990

[^4]:    ${ }^{2}$ M. Capek, V. Losenicky, L. Jelinek, et al., "Validating the characteristic modes solvers", IEEE Trans. Antennas Propag., vol. 65, no. 8, pp. 4134-4145, 2017. DOI: 10.1109/TAP. 2017. 2708094

[^5]:    ${ }^{3}$ G. Kristensson, Scattering of electromagnetic waves by obstacles. Edison, NJ: SciTech Publishing, an imprint of the IET, 2016

[^6]:    ${ }^{3}$ G. Kristensson, Scattering of electromagnetic waves by obstacles. Edison, NJ: SciTech Publishing, an imprint of the IET, 2016

[^7]:    ${ }^{3}$ G. Kristensson, Scattering of electromagnetic waves by obstacles. Edison, NJ: SciTech Publishing, an imprint of the IET, 2016

[^8]:    ${ }^{3}$ G. Kristensson, Scattering of electromagnetic waves by obstacles. Edison, NJ: SciTech Publishing, an imprint of the IET, 2016

[^9]:    ${ }^{4}$ D. Tayli, M. Capek, L. Akrou, et al., "Accurate and efficient evaluation of characteristic modes", , 2017, submitted, arXiv:1709.09976. [Online]. Available: https://arxiv.org/abs/1709.09976

[^10]:    ${ }^{4}$ D. Tayli, M. Capek, L. Akrou, et al., "Accurate and efficient evaluation of characteristic modes", , 2017, submitted, arXiv:1709.09976. [Online]. Available: https://arxiv.org/abs/1709.09976

[^11]:    ${ }^{4}$ D. Tayli, M. Capek, L. Akrou, et al., "Accurate and efficient evaluation of characteristic modes", , 2017, submitted, arXiv:1709.09976. [Online]. Available: https://arxiv.org/abs/1709.09976

[^12]:    ${ }^{5}$ C. G. Montgomery, R. H. Dicke, and E. M. Purcell, Principles of microwave circuits. New York, United States: McGraw-Hill, 1948
    R. J. Garbacz, "A generalized expansion for radiated and scattered fields", PhD thesis, Department of Electrical Engineering, Ohio State University, 1968

